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### A 64 × 64 SPAD-Based Indirect Time-of-Flight Image Sensor with a depth range of 50m

International SPAD Sensor Workshop 2022

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# Outline

- Motivation SPAD-based Time-of-Flight sensor
- Proposed SPAD-based iToF sensor
  - Pulse-shaping circuit
  - Analog pulse counter
- Measurement results
- Conclusion



# 2 types of ToF sensors





S. W. Hutchings [JSCC' 19]

- Time-of-Flight (ToF) Sensors
  - Direct: Long-range (> 100m) ☺, large pixel pitch, 3-D stacking ⊗
  - Indirect: High pixel resolution, high frame rate 🙂, short-range 😕



#### **Trend in ToF sensors**



M.-S. Keel [ISSCC' 21]



A. R. Ximenes [JSSC' 19]

- Multiple integration capacitor (iToF), up to 4 taps
- Small pixel pitch, down to 3.5 µm (iToF), 9.2 µm (dToF)
- 3-D stacked wafer (iToF, dToF both)
  - Resulting in a high production cost ⊗



- Average of 35 m is required for emergency braking (at 100 km/h)
- Required distance: 50 m, FoV: 20 °
  - Pixel resolution: at least 60 (to detect pedestrian)
  - iToF sensor: high frame rate, high accuracy, pixel scalability√

# **Principle of indirect ToF**



- Correlation between reflected light and time-window for iToF sensors
  - Integrate photon current in selected time-window (PD)
  - Counting photon in selected time-window (SPAD)
- Immune to background light <sup>(c)</sup> (removed as an offset)



- PW / DNW junction SPAD with retrograde DNW guard-ring
  - Rounded corner shape > maximize active area
  - Integrated DNW > thin cathode (1  $\mu$ m)
  - Improved fill factor > up to 50 % (previous work, SSC-L' 20)

#### Time-gated photon counting V<sub>SPAD</sub> Photon SPAD Photon Vs Vs $\phi_{TRG}$ $\phi_{TW}$ Clk $\phi_{TW}$ (ON) V<sub>PR</sub> **0-**Counter NCOUNT Q NCOUNT Pulse-shaping circuit

- Logic gate (pulse-shaping circuit) is attached for gating V<sub>S</sub>
  - $V_{S}$  propagates to counter only  $\Phi_{TW}$  is on (enable correlation)
- However, passive recharge slows down counting rate > requires active recharge circuit



# Implementing pulse-shaping circuit



- 2 inverters are added to sharpen pulse edge of V<sub>S</sub>
  - First inverter: skewed between P/N for threshold control
  - Second inverter: added for the polarity of the  $\Phi_{AR}$
- Pulse-shaping circuit: inverter outputs and "NOR gate"



# Implementing pulse-shaping circuit



- 3-input NOR gate: gated by  $\Phi_{TW}$  for time-gated photon counting
  - Operation rate of pulse counter: determined by feedback delay
- Multiple counter: motion artifact↓, light efficiency↑



#### Selecting pulse counter



**Digital counter** 

Analog counter

- Digital / analog pulse counter
  - Digital : high speed, linearity, no additional readout circuit large pixel pitch, 3-D stack process 8
  - Analog : area-efficient, pixel-level implementation 
     medium operation speed, limited linearity 
     Image: Speed and Speed and

#### Charge-sharing based counter





- Charge sharing between integration (C<sub>I</sub>) & degeneration capacitor (C<sub>D</sub>)
- 2 NMOS transistors used as switches ( $\Phi_{TRIG}$  and  $\Phi_{AR}$ )
  - Counting step,  $\Delta V_{OUT} \propto V_{OUT}$  > limited linearity & no tunability
- Additional technique for improving linearity

#### **Charge-injection based counter**





- Source follower ( $M_F$ ) is added, separating  $C_D$  from  $V_{OUT}$ 
  - Constant charge-injection and offers tunability by changing V<sub>SF</sub>
  - Suffers threshold variation, resulting in counting step variation

#### **Proposed analog counter**





- Amplifier regulates source voltage of M<sub>F</sub> to V<sub>REF</sub>
  - Constant charge-injection without suffering threshold variation
  - Offers step tunability by adjusting V<sub>REF</sub>

#### **Pulse counter comparison**



Structure	Digital	Charge- sharing	Charge- injection	Proposed
Occupied area [µm <sup>2</sup> ]	1200	$40^{\dagger}$	44†	60†
Simulated linearity	-	6 bit	8 bit	9 bit
Tunability	Х	Х	0	0

<sup>†</sup>Estimated value with  $C_I = 250$  fF

- Digital counter occupies largest area
- Analog counter shows limited linearity (< 8-bit)
  - Charge-injection based counter may suffer threshold variation
- Proposed counter achieves high linearity (> 9-bit), tunability <sup>(C)</sup>



#### **Proposed counter: Discharge phase**



- Amplifier: 5 transistor with 600 nA bias, 30 dB DC gain
- $M_{F1}$ ,  $M_{F2}$ : NMOS switches ,  $C_P$ : parasitic capacitor at  $V_P$ 
  - $C_P$ : charged from ground to  $V_{REF}$ ,  $C_I$ : discharged by  $Q_P$  ( $\Phi_{TRIG}$ )
  - Counting step  $\Delta V_0 = C_P / C_I * V_{REF}$



#### **Proposed counter: Reset phase**



- $C_P$  is reset to ground ( $\Phi_{DIS}$ ), remaining still until next trigger
  - Discharge and reset phases: should be separated
  - Requires delay between control signals ( $\Phi_{TRIG}$  and  $\Phi_{DIS}$ )

# Schematic of the proposed pixel



- AND gate: adding delay between  $\Phi_{AQ}$  and  $\Phi_{AR}$ 
  - Disabling SPAD:  $\Phi_{AQ}$  to low ( $V_{PIX} > V_{EX}$ )
  - M<sub>AR</sub>: turned off to avoid short circuit current √
- PMOS source follower & row switch: pixel readout

#### Layout of the proposed pixel





- Pulse shaping: 8.5 µm, counter: 20.5 µm length
  - Layout optimization: C<sub>IA</sub> and C<sub>IB</sub>
  - Source follower: below column readout line



### Chip micrograph



- 110 nm FSI process, core area: 3 x 3 mm<sup>2</sup>
  - 64 x 64 SPAD pixel array
  - 256 single-slope ADC (4 ADCs / column)



- Global reset, pixel integration: 125 µs, pixel readout: 1.8 ms
- Frame rate of the proposed sensor: 520 fps (2-D)
  - 3-D frame: consists of four 2-D frames
  - Two 3-D frames (high, low  $f_{demod}$ ) > 65 fps (depth image)



- VCSEL (850 nm) emission power: 1.16 mW/cm<sup>2</sup> (@ 1 m)
  - VCSEL emission angle: 20 °, FoV of optical lens: 17 °
  - Optical filter > CWL: 850 nm, FWHM: 10 nm, OD: 4.0
- Power consumption: 42.7 mW



- 64 x 64 SPAD array dark count rate (DCR)
  - 33 (median) / 1800 cps (mean)
- Photon detection probability (PDP)
  - Peak: 28.2 % (@ 480 nm), 5.85 % (@ 850 nm)

# **Comparison with other SPADs**



- Proposed SPAD and SPADs in similar technology nodes
  - Achieves lowest normalized DCR and comparable PDP
  - Low excess bias voltage, deep junction (PW/DNW)

### Step tunability of analog counter





- Measured counting step versus V<sub>REF</sub>
  - $V_{REF}$  tuned from 0.2 to 1 V, counting step > 0.6 to 3 mV
  - $V_{REF} = 0.5 \text{ V}, C_P = 0.75 \text{ fF}, C_I = 250 \text{ fF} > 9-\text{bit counter}$



- Measured noise and step variation of proposed counter

  - Measured noise  $(1-\sigma)$ : > 0.7 LSB for entire codes
  - Step variation: 8.3 %, offset mismatch and C<sub>P</sub> variation



- Measured DNL / INL of analog pulse counter array
  - DNL: +0.25 / -0.19 LSB, INL: +0.22 / -0.72 LSB
  - Outliers: less than 3 % of pixel population
  - Proposed pulse counter: operates as 9-bit counter 🙂



- Measured depth uncertainty  $(1-\sigma)$  result in range of 5 to 50 m (65 fps)
  - Depth uncertainty: STD of distance over 400 measurements
  - 1.35 to 11.3 cm in range of 5 to 50 m
  - Relative depth uncertainty: 0.22 % (at 50 m)

# Depth uncertainty result with BGL





- Measured depth uncertainty varying background-to-signal ratio (BSR)
  - Variable IR source: applied as background (target @ 1.5 m)
  - Depth uncertainty increases until counter saturation
  - Accumulating multiple frames improves depth uncertainty

#### **3-D measurement results**



 $f_{demod}$  = 50 MHz

- Sample 3-D images of Agrippa statue
  - Demodulated with maximum designed frequency (50 MHz)
  - Result shows detail with high demodulation frequency

# **3-D** measurement results Phase (degree) -90 -180

BGL = 120 klx





- Sample 3-D images with 120 klx sunlight
  - Vertical wall: 20 m, 8-frame accumulation (16.25 fps)
  - Optical bandpass filter: reduce sunlight by 95 %



- Depth uncertainty improvement by frame accumulation
  - Increasing frames by 4 times > more details & lower noise
  - Optical bandpass filter & frame accumulation

> 3-D image is reconstructed successfully!

	This Work	B. Park	C. S. Bamji	T. Okino	E. Manuzzato	M. Perenzoni	C. Niclass	S. W	
		SSC-L20 [1]	ISSCC18 [2]	ISSCC20 [3]	ISSCC22 [4]	SSC-L20 [5]	JSSC14 [6]	JS	
Photodetector	SPAD	SPAD	PD	SPAD (Vertical)	SPAD	SPAD	SPAD	STAD	SI AD
ToF technique	iToF	iToF	iToF	iToF + dToF	dToF	dToF	dToF	dToF	dToF
Key feature	2-tap pulse counter	1-tap pulse counter	2-tap photogate	1-tap pulse counter	Per-pixel TDC	Per-column TDC	Per-column TDC	16-Shared TDC	128-Shared TDC
Technology [nm]	110	110	65 (BSI)	65	110	150	180	40 / 90	45 / 65
Pixel resolution	64 × 64	64 × 64	1024 × 1024	1200 × 900	64 × 64	50 × 40 (2-SPAD)	202 × 96	256 × 256 (4 × 4 SPAD)	256 × 256 (16 × 8 SPAD)
Pixel pitch [µm]	32	32	3.5	6	48	38.5 × 33.5	N/A	9.2 (36.8 × 36.8)	19.8 (316.8 × 158.4)
Fill factor [%]	26.3	17.3	100 (with μ-lens)	N/A	12.9	4.8-15.3	70	51	31.3
Background [klx]	<b>120</b> <sup>†</sup>	$90^{\dagger}$	25	N/A	30	18	$100^{\dagger}$	1	3
Emitter wavelength [nm]	850	850	860	N/A	905	650	870	671	532
Frame rate [fps]	65	90	30	30	25	1 kHz @ 1000 pts	10	30	2000 pixel/s
Maximum distance [m]	50	40	4.2	13(iToF) / 250(dToF)	8.2	3	100	50	150
Depth uncertainty [%]	0.22	0.51	0.2	0.38 / N/A	3.29	0.05	0.14	N/A	0.1
Chip power [mW]	42.7	33.5	650	2500	205.74	28.3	530	77.6	N/A

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# Conclusion

- 64 x 64 SPAD-based iToF sensor with 2-tap analog pulse
- Time-gated photon counting method
  - Compact pixel pitch (32 µm), high fill factor (26.3 %)
  - Large detection range (50 m), low depth uncertainty (0.22 %)
  - High frame rate (65 fps), high sunlight tolerance (120 klx)
    > Suitable for outdoor applications
- This work shows high potential of SPAD-based iToF sensors



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# Thank you for your attention !